

motor_fft_cpu.cpp

Complete Technical Reference

Motor Fault Detection & Full Harmonic Analysis — C++23

Language	C++23
Sample rate	32 000 Hz
FFT size	65 536 pts (fixed)
Channels	7 (IA, VA, IB, VB, IC, VC, IN)
Frame size	3 610 bytes/frame, 128 samples/frame
Output modes	Human-readable / JSON
Dependencies	Standard library only (<code>-lm</code>)

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Overview

`motor_fft_cpu.cpp` is a standalone C++23 binary that performs Motor Current Signature Analysis (MCSA) entirely on the CPU without any external DSP library. It reads a raw multi-channel binary capture file (`tcp.raw`), computes a Cooley–Tukey FFT on each of the seven acquisition channels, and produces a rich fault-detection report either as human-readable text or as a structured JSON object suitable for ingestion by downstream tools such as `diagnose.py` and MongoDB.

The program implements the following analytical layers:

- **Global FFT** with Hann windowing (65 536-point Cooley–Tukey, $\Delta f \approx 0.488$ Hz/bin).
- **Harmonic extraction** via least-squares sinusoidal fit for harmonics H_1 – H_{20} on all channels.
- **Broken-rotor-bar** sideband detection at $f_0(1 \pm 2s)$.
- **Eccentricity** sideband detection at $f_0 \pm kf_r$, $k = 1, 2, 3$ and high-order families at $3f_0 \pm f_r$, $5f_0 \pm f_r$.
- **Bearing fault** detection (BPFO, BPFI, BSF, FTF).
- **Mechanical faults**: rotor unbalance, misalignment, looseness, blade pass (pump impeller), and cavitation.
- **STFT / wavelet / synchrosqueezed transform** advanced layer.
- **NI Peak Detector VI** equivalent with sub-sample interpolation.
- **Fault hypothesis scoring** with multi-domain evidence fusion.

Build & Compilation

The source file has no external dependencies beyond the C standard math library. It compiles with either **Clang** or **GCC**:

```
# Clang (recommended for best C++23 support)
clang++ -std=c++23 -O3 -o motor_fft_cpu motor_fft_cpu.cpp -lm

# GCC
g++ -std=c++23 -O3 -o motor_fft_cpu motor_fft_cpu.cpp -lm
```

The `-O3` flag is important: the FFT inner loop benefits significantly from auto-vectorisation.

Inputs

Command-Line Interface

```
./motor_fft_cpu <input.raw> [f0] [rpm] [poles] [OPTIONS...]
```

Positional arguments

Position	Name	Default	Description
1	<code>input</code>	— (<i>required</i>)	Path to the raw binary capture file (<code>.raw</code>) produced by the TCP acquisition board.
2	<code>f0</code>	60 Hz	Nominal supply frequency in hertz. Used as the prior for automatic f_0 detection.
3	<code>rpm</code>	1785 RPM	Nameplate motor speed. Used to compute slip and all sideband frequencies.
4	<code>poles</code>	4	Number of magnetic poles. Used to compute synchronous speed $n_s = 120f_0/p$.

Parameter priority: positional arguments > `-dsp dsp.json` > internal defaults.

Named options

Option	Description
<code>-json</code>	Switch output to JSON mode. Required for integration with <code>diagnose.py</code> and MongoDB storage.
<code>-dsp dsp.json</code>	Load f_0 , RPM, and poles automatically from a DSP runtime JSON file. The program reads keys <code>"frequency_hz"</code> , <code>"motor.rpm"</code> , and <code>"motor.poles"</code> . VFD motors are detected via <code>"motor.is_vfd"</code> .
<code>-cal calibration.json</code>	Apply per-channel physical calibration gains (A/code or V/code) to convert raw ADC codes to amperes and volts. Without this flag all amplitudes are expressed as % of full scale (%FS).
<code>-bearings bearings.json</code>	Load external bearing geometry for runtime computation of BPFO, BPFI, BSF, and FTF. Without this flag the program falls back to hardcoded SKF6205 ratios.
<code>-ref {IA IB IC}</code>	Choose the reference current channel for fault detection. Default: IA.
<code>-spectrum <mode></code>	Specify which spectra are included in the JSON output. Valid modes: IA, IB, IC, IN, VA, VB, VC, ALL_CURRENT, ALL_VOLTAGE, ALL.
<code>-include-voltage-spectrum</code>	Enable voltage spectrum export in JSON. Required for voltage channels to appear even when <code>-spectrum ALL</code> is specified.
<code>-max-freq N</code>	Truncate the serialised spectrum in the JSON to N Hz. Does not affect the internal FFT computation or fault detection, which always use the full bandwidth to Nyquist.
<code>-no-autodetect-f0</code>	Disable spectral f_0 auto-detection. The program uses the provided or default f_0 value exactly.

Input File Format: `tcp.raw`

The raw capture file is a stream of fixed-size **frames**. Each frame has the following binary layout:

Field	Bytes	Description
Header	26	Frame header (timestamp, sequence number, etc.). Skipped by the decoder.
Payload	3584	$128 \times 7 \times 4 = 3584$ bytes of interleaved 32-bit signed integers (one sample per channel per set).
Total	3610	bytes per frame

The seven channels are interleaved in the order:

$$IA = 0, VA = 1, IB = 2, VB = 3, IC = 4, VC = 5, IN = 6$$

Each 32-bit signed integer is a raw ADC code. Conversion to physical fraction of full scale is:

$$x_{FS} = \frac{\text{code}}{2^{23}} = \text{code} \times 1.192 \times 10^{-7} \quad (1)$$

When a calibration file is supplied the per-channel gain g_c (in A/code or V/code) replaces the above default:

$$x_{\text{phys}} = \text{code} \times g_c \quad (2)$$

Constants defined in source:

- `FRAME_BYTES = 3610`
- `HEADER_BYTES = 26`
- `PAYLOAD_BYTES = 3584`
- `SETS_PER_FRAME = 128`
- `N_CH = 7`
- `FS_HZ = 32000`
- `NORM_FACTOR = 1/8,388,608  $\approx 1.192 \times 10^{-7}$`

Configuration Files (JSON inputs)

dsp.json — DSP runtime configuration

Loaded with `-dsp`. The program reads the following fields:

```
{
  "frequency_hz": 60.0,
  "motor": {
    "rpm": 1785.0,
    "poles": 4,
    "is_vfd": false,
    "vfd_freq_hz": 0.0
  }
}
```

Key	Type	Description
frequency_hz	float	Grid/supply frequency measured by the ADE9000 power metering IC. For VFD motors this is the <i>grid</i> frequency, not the motor current fundamental.
motor.rpm	float	Motor shaft speed in RPM (from nameplate or encoder).
motor.poles	integer	Number of magnetic poles.
motor.is_vfd	boolean	Set <code>true</code> when the motor is driven by a Variable Frequency Drive. The f_0 auto-detection search range is expanded to [5, 200] Hz.
motor.vfd_freq_hz	float	Last-known VFD output frequency (may be stale). Used as the search hint for VFD f_0 detection.

calibration.json — Physical calibration gains

```
{
  "current_gain": { "a": 0.003052, "b": 0.003048,
                   "c": 0.003055, "n": 0.003052 },
  "voltage_gain": { "a": 0.12207,  "b": 0.12200,
                   "c": 0.12210 }
}
```

All gain values are in **A/code** (current) or **V/code** (voltage). If any gain is missing or zero the program falls back to %FS units.

bearings.json — External bearing geometry

```
{
  "model": "WeatherfordMG_pump",
  "bearings": [
    {
      "id": "DE",
      "location": "drive_end",
      "d_mm": 25.0,
      "D_mm": 52.0,
      "Dw_mm": 7.938,
      "Z": 9,
      "alpha_deg": 0.0,
      "critical": true,
      "note": "SKF 6205-Z equivalent"
    }
  ],
  "analysis_index": 0,
  "impeller_blades": 7
}
```

Key	Type	Description
id	string	Identifier for this bearing (e.g. "DE", "NDE").
location	string	Human-readable location label.
d_mm	float	Bearing bore diameter (inner race diameter) [mm].
D_mm	float	Outer race diameter [mm].
Dw_mm	float	Rolling element (ball/roller) diameter [mm].
Z	integer	Number of rolling elements.
alpha_deg	float	Contact angle in degrees (0 for deep-groove ball bearings).
critical	boolean	Mark as safety-critical bearing.
analysis_index	integer	Index into <code>bearings[]</code> array to use for fault frequency calculation.
impeller_blades	integer	Number of pump impeller blades. Enables blade-pass fault detection at $f_0 \pm Z_b f_r$.

Signal Processing Pipeline

Step 1: Data Decoding

The program reads `tcp.raw` frame-by-frame. For each frame the 26-byte header is skipped and the 3584-byte payload is unpacked as 128×7 interleaved 32-bit signed integers. Each integer is multiplied by the per-channel gain to produce a floating-point signal vector $x_c[n]$ for channel c .

Step 2: FFT Size Selection

The FFT length N is chosen as follows:

$$N = \min(N_{\text{fixed}}, 2^{\lceil \log_2 L \rceil}), \quad N_{\text{fixed}} = 65536 \quad (3)$$

where L is the total number of samples available. For captures shorter than $N_{\text{fixed}}/f_s = 2.048$ s a smaller power-of-two is used automatically via `next_pow2(L)`.

The frequency resolution is:

$$\Delta f = \frac{f_s}{N} = \frac{32000}{65536} \approx 0.4883 \text{ Hz/bin} \quad (4)$$

Step 3: Hann Windowing

Each segment is multiplied sample-by-sample by a Hann window to reduce spectral leakage. The window function is:

$$w[n] = \frac{1}{2} \left(1 - \cos \frac{2\pi n}{N-1} \right), \quad n = 0, 1, \dots, N-1 \quad (5)$$

The coherent gain C_g of the Hann window is:

$$C_g = \frac{1}{N} \sum_{n=0}^{N-1} w[n] = 0.5 \quad (6)$$

Peak amplitudes extracted from the FFT are divided by C_g to recover calibrated physical amplitudes.

Step 4: Cooley–Tukey Radix-2 FFT

The program includes a self-contained in-place radix-2 Cooley–Tukey FFT with no external library dependency. The algorithm operates on a complex array $\mathbf{X} \in \mathbb{C}^N$:

Bit-reversal permutation

Indices are permuted by bit-reversal before the butterfly passes. For an N -point transform the permutation maps index i to the integer obtained by reversing the $\log_2 N$ binary bits of i .

Butterfly stages

After bit-reversal, $\log_2 N$ butterfly stages are applied:

$$X[i+j] \leftarrow X[i+j] + W^j \cdot X[i+j+\frac{\ell}{2}] \quad (7)$$

$$X[i+j+\frac{\ell}{2}] \leftarrow X[i+j] - W^j \cdot X[i+j+\frac{\ell}{2}] \quad (8)$$

where $\ell \in \{2, 4, 8, \dots, N\}$ is the current stage length and the twiddle factor is:

$$W = e^{-j2\pi/\ell} \quad (9)$$

This gives the standard DFT result:

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1 \quad (10)$$

with time complexity $\mathcal{O}(N \log_2 N)$.

Step 5: Magnitude Spectrum

The single-sided (positive-frequency) magnitude spectrum is:

$$|X[k]| = \sqrt{\text{Re}(X[k])^2 + \text{Im}(X[k])^2}, \quad k = 0, 1, \dots, \frac{N}{2} \quad (11)$$

The peak amplitude for harmonic h at bin k_h (after coherent-gain correction) is:

$$A_h = \frac{2}{N \cdot C_g} |X[k_h]| \quad (12)$$

(factor 2 accounts for the one-sided spectrum; factor $1/(N \cdot C_g)$ normalises for window and FFT scaling.)

Step 6: Fundamental Frequency Auto-Detection

Unless `-no-autodetect-f0` is used, the program refines f_0 by searching for the strongest bin in a window around the nominal value:

- **Normal motor:** search window $[f_{0,\text{prior}} \times 0.95, f_{0,\text{prior}} \times 1.05]$ (5% tolerance).
- **VFD motor** (`is_vfd=true`): search window $[5, 200]$ Hz (full VFD range). Uses `vfd_freq_hz` as the search hint if available.

The detected peak is accepted only if the deviation from the prior satisfies $|f_{0,\text{detected}} - f_{0,\text{prior}}| \leq \delta$ where $\delta = \max(0.5 \text{ Hz}, 0.05 f_{0,\text{prior}})$.

Step 7: Harmonic Extraction (Least-Squares Sinusoidal Fit)

For each harmonic $h = 1, \dots, 20$ at target frequency $f_h = h f_0$, the program performs a least-squares fit of the form:

$$\hat{x}[n] = A_h \cos(2\pi f_h n / f_s) + B_h \sin(2\pi f_h n / f_s) \quad (13)$$

over the windowed segment. The amplitude and phase are:

$$\hat{A}_h = \sqrt{A_h^2 + B_h^2} \quad (14)$$

$$\phi_h = \arctan_2(B_h, A_h) \quad (15)$$

The RMS value of each harmonic is $A_{h,\text{rms}} = \hat{A}_h / \sqrt{2}$.

Total harmonic distortion (THD) is defined as:

$$\text{THD} = \frac{\sqrt{\sum_{h=2}^{20} A_{h,\text{rms}}^2}}{A_{1,\text{rms}}} \times 100\% \quad (16)$$

Motor Physics Model

Fundamental Mechanical Quantities

$$f_r = \frac{n_{\text{rpm}}}{60} \quad [\text{Hz}] \quad (17)$$

$$n_s = \frac{120 f_0}{p} \quad [\text{RPM, synchronous speed}] \quad (18)$$

$$s = \frac{n_s - n_{\text{rpm}}}{n_s} \quad [\text{per-unit slip}] \quad (19)$$

where p is the number of poles.

Slip is validated to $0 < s < 0.20$. If slip falls outside this range the physics model is marked invalid and sideband detection is skipped.

Broken-Rotor-Bar Sideband Frequencies

MCSA theory predicts that a broken rotor bar produces current sidebands symmetrically around the fundamental at:

$$f_{BB} = f_0 (1 \pm 2s) \quad (20)$$

These are the **most important** fault frequencies for MCSA. The sidebands are symmetric around f_0 which distinguishes them from eccentricity sidebands.

Eccentricity Sideband Frequencies

Static or dynamic air-gap eccentricity produces sidebands at:

$$f_{ecc,k}^{\pm} = f_0 \pm k f_r, \quad k = 1, 2, 3 \quad (21)$$

High-order families at the 3rd and 5th harmonics:

$$f_{ecc,hf}^{\pm} = m f_0 \pm f_r, \quad m \in \{3, 5\} \quad (22)$$

Bearing Fault Frequencies

The four fundamental bearing defect frequencies are derived from the bearing geometry. Let:

- $P_d = (d + D)/2$ — pitch diameter [mm]
- D_w — rolling element diameter [mm]
- Z — number of rolling elements
- α — contact angle [degrees]
- f_r — mechanical rotation frequency [Hz]

Define the ratio:

$$\rho = \frac{D_w}{P_d} \cos \alpha \quad (23)$$

Then:

$$\text{FTF} = \frac{f_r}{2} (1 - \rho) \quad (24)$$

$$\text{BPFO} = \frac{Z f_r}{2} (1 - \rho) \quad (25)$$

$$\text{BPFI} = \frac{Z f_r}{2} (1 + \rho) \quad (26)$$

$$\text{BSF} = \frac{P_d f_r}{2 D_w} (1 - \rho^2) \quad (27)$$

When no external bearing file is supplied the program falls back to hardcoded SKF6205 ratios (relative to f_r): BPFO= 3.585, BPFI= 5.415, BSF= 2.357, FTF= 0.398.

Additional Mechanical Fault Frequencies

$$f_{\text{unbal}}^{\pm} = f_0 \pm f_r \quad (28)$$

$$f_{\text{misalign},k}^{\pm} = f_0 \pm k f_r, \quad k = 1, 2, 3 \quad (29)$$

$$f_{\text{loose},n}^{\pm} = f_0 \pm n \frac{f_r}{2}, \quad n = 1, 2, 3 \quad (30)$$

$$f_{\text{blade}}^{\pm} = f_0 \pm Z_b f_r \quad (31)$$

where Z_b is the number of impeller blades (configured via `bearings.json`).

Cavitation does not produce discrete frequency lines; instead, it elevates the broadband energy in the 5–30 Hz band. The program computes:

$$E_{\text{cav}} = 20 \log_{10} \left(\frac{1}{B} \sqrt{\sum_{f=5}^{30} |X(f)|^2} / A_{1,\text{rms}} \right) [\text{dB}] \quad (32)$$

where B is the number of bins in the 5–30 Hz band.

Advanced Analysis Layer

Short-Time Fourier Transform (STFT)

The STFT segments the reference channel into overlapping windows and applies the FFT to each, producing a time-frequency matrix $S[t, f]$ (magnitude in dB). This enables detection of non-stationary fault signatures that would be averaged out in a single global FFT.

Wavelet Packet Transform (WPT)

The program applies two wavelet families:

- **Haar (db1)**: simple, fast, orthogonal.
- **DB4 (Daubechies-4)**: superior frequency localisation for MCSA.

For each decomposition level ℓ and node n , the normalised energy is:

$$E_{\ell,n} = \frac{\sum_k d_{\ell,n}[k]^2}{\sum_{\ell',n'} \sum_k d_{\ell',n'}[k]^2} \quad (33)$$

WPT nodes are labelled with physical frequency ranges to provide interpretable zone energy (e.g. *fundamental*, *bearing_zone*, *slot_harmonic*).

Synchrosqueezed STFT (SST)

The synchrosqueezed transform reallocates STFT energy along instantaneous frequency ridges. This provides sharper time-frequency localisation than the standard STFT, which is beneficial for detecting slowly varying f_0 or slip in VFD-driven motors.

V/I Power Analysis

For each STFT window the program computes power-quality quantities at the fundamental frequency:

$$P_1 = \text{Re}(V_1 \cdot I_1^*) \quad (34)$$

$$Q_1 = \text{Im}(V_1 \cdot I_1^*) \quad (35)$$

$$\text{PF}_1 = \frac{P_1}{|V_1| |I_1|} \quad (36)$$

$$Z_1 = \frac{V_1}{I_1} \quad (37)$$

NI Peak Detector VI Equivalent

The NI Peak Detector replicates the algorithm of the LabVIEW *NI_AAL_SigProc.lvlib:Peak Detector.vi*. A quadratic polynomial is fitted by least-squares over a window of `width` samples centred on each local maximum:

$$\hat{y}(x) = a_0 + a_1x + a_2x^2 \quad (38)$$

The symmetric normal equations (odd moments vanish):

$$a_1 = \frac{S_{xy}}{S_2} \quad (39)$$

$$a_0 = \frac{S_4S_y - S_2S_{x^2y}}{S_0S_4 - S_2^2} \quad (40)$$

$$a_2 = \frac{S_0S_{x^2y} - S_2S_y}{S_0S_4 - S_2^2} \quad (41)$$

Sub-sample peak location (fractional bin) and fitted amplitude:

$$\hat{n} = n_0 - \frac{a_1}{2a_2} \quad (42)$$

$$\hat{A} = a_0 - \frac{a_1^2}{4a_2} \quad (43)$$

Peaks are accepted only when $\hat{A}$ exceeds the threshold *and* the local SNR ≥ 6 dB above the local noise floor.

Slip Estimation from Spectrum

The motor slip is estimated directly from the broken-bar sideband positions in the current spectrum (no external tachometer required):

1. Scan $[f_0(1 - 2s_{\max}), f_0(1 - 2s_{\min})]$ for the strongest peak.
2. Scan $[f_0(1 + 2s_{\min}), f_0(1 + 2s_{\max})]$ for the strongest peak.
3. Require each peak > 6 dB above local noise floor.

4. Check symmetry: $|f_0 - f_{lo}| \approx |f_{hi} - f_0|$.
5. Estimate slip: $\hat{s} = (f_0 - f_{lo}) / (2 f_0)$ (SNR-weighted if both visible).
6. Validate $s_{\min} < \hat{s} < 0.15$.

Fault Hypothesis Scoring

After all spectral and wavelet quantities are computed, the program fuses evidence into probabilistic hypothesis scores $h \in [0, 1]$ for five fault classes.

Score Functions

A linear ramp between a warning threshold T_w and a saturation threshold T_s :

$$\text{score_thresh}(x, T_w, T_s) = \text{clamp}\left(\frac{x - T_w}{T_s - T_w}, 0, 1\right) \quad (44)$$

A soft compression to avoid hard saturation:

$$\text{compress}(x, \gamma) = \frac{x}{x + \gamma^{-1}(1 - x)} \quad (45)$$

Broken-Bar Hypothesis

$$h_{\text{BB}} = \text{clamp}\left(\underbrace{C(g_{\text{BB}}, 1.8)}_{\text{sideband level}} \cdot \underbrace{N_{\text{BB}}}_{\text{noise gate}} \cdot \underbrace{(0.7 + 0.3 \text{ sym})}_{\text{symmetry}} \cdot \underbrace{(0.7 + 0.3 \text{ slip_cons})}_{\text{slip consistency}}, 0, 1\right) \quad (46)$$

where $C(g, \gamma)$ denotes the score compressed with exponent γ , N_{BB} is the minimum noise floor score of the two sidebands, and sym is the sideband symmetry score $\min(r_{lo}, r_{hi}) / \max(r_{lo}, r_{hi})$.

Eccentricity Hypothesis

$$h_{\text{ecc}} = \text{clamp}\left(C(g_{\text{ecc}}, 1.6) \cdot N_{\text{ecc}}, 0, 1\right) \quad (47)$$

Bearing Hypothesis

$$h_{\text{brg}} = \text{clamp}\left(C(g_{\text{brg}}, 1.6) \cdot N_{\text{brg}}, 0, 1\right) \quad (48)$$

Supply Harmonic H3 Hypothesis

$$h_{\text{H3}} = \begin{cases} C(g_{\text{H3}}, 1.5) & \text{if } H_{3,I} > 0.015 \wedge H_{3,V} > 0.0075 \wedge H_{3,V} \geq 0.35 H_{3,I} \\ 0 & \text{otherwise} \end{cases} \quad (49)$$

Fault Competition (Post-Processing)

A competition step prevents simultaneous broken-bar and eccentricity classification when evidence is ambiguous. The sideband asymmetry score is computed as $1 - \text{sym}$ to distinguish eccentricity (asymmetric sidebands) from broken-bar (symmetric sidebands).

Binary Fault Flags

The final binary fault flags apply additional hard gates on top of the hypothesis scores. For broken-bar:

fault_broken_bar is true if and only if all of the following hold:

- `physics_valid & slip_valid`
- Sidebands confirmed in ≥ 2 of 3 phases
- $\min(r_{lo}, r_{hi}) \geq \text{dynamic ratio gate}$
- Noise score ≥ 0.35
- $h_{BB} \geq 0.35$ **and** ($h_{BB} \geq 1.1 h_{ecc}$ or $h_{BB} \geq 0.60$)

Outputs

Human-Readable Mode (default)

Without `-json` the program prints a plain-text report to `stdout` with the following sections:

```
[FFT] IA: f0=60.00 Hz  H1=12.453 A  THD=2.31%  RMS=12.490 A
[FFT] IB: f0=60.00 Hz  H1=12.421 A  THD=2.29%  RMS=12.458 A
[FFT] IC: f0=60.00 Hz  H1=12.477 A  THD=2.34%  RMS=12.514 A
...
[MOTOR] f0=60.00 Hz  fmec=29.75 Hz  slip=0.0083  rpm=1785.0
[BB]   lo=59.003 Hz  ratio=0.00121  hi=60.997 Hz  ratio=0.00118
[ECC]  k=1  lo=30.25 Hz  ratio=0.00042  hi=89.75 Hz  ratio=0.00039
[BPFO] 106.6 Hz  ratio=0.000081
...
FAULT: none
```

Diagnostic messages (`[DSP]`, `[CAL]`, `[f0]`, `[FMEC]`, `[SLIP]`, `[NIPEAK]`) are written to `stderr`.

JSON Mode (`-json`)

The full output is a single JSON object written to `stdout`. The top-level schema is described below. All JSON text uses Courier/console font in green on a light background.

Top-level structure

```
{
  "schema_version": 2,
  "source_file":    "tcp.raw",
  "ref_channel":    "IA",
  "calibrated":     true,
  "n_frames":       512,
  "fft_n":          65536,
  "df_hz":          0.488281,
  "f0_prior":       60.0,
  "f0_used":        59.998,
  "f0_auto":        true,
  "f0_source":      "autodetect_refined",
  "f0_from_dsp_hz": 60.0,
  "autodetect_accepted": true,
```

```

"motor": { ... },
"faults": { ... },
"current": [ {}, {}, {}, {} ],
"voltage": [ {}, {}, {} ],
"advanced": { ... },
"decision": { ... }
}

```

motor object

```

"motor": {
  "f0_hz":      59.998,
  "rpm":        1785.3,
  "fmec_hz":    29.755,
  "slip":        0.00822,
  "poles":      4,
  "physics_valid": true,
  "slip_valid":  true,
  "bb_lo_hz":   58.010,
  "bb_hi_hz":   61.986,
  "ecc_l_lo_hz": 30.243,
  "ecc_l_hi_hz": 89.753,
  "bpfo_hz":    106.62,
  "bpfi_hz":    161.10,
  "bsf_hz":     70.13,
  "ftf_hz":     11.85,
  "bearing_model": "SKF6205_fallback"
}

```

faults object

```

"faults": {
  "broken_bar":      false,
  "eccentricity":     false,
  "bearing":          false,
  "thd_warn":         false,
  "thd_crit":         false,
  "h3":               false,
  "h3_from_grid":     false,
  "unbalance":        false,
  "rotor_unbalance":  false,
  "misalignment":     false,
  "looseness":        false,
  "blade_pass":       false,
  "cavitation":       false,
  "unbalance_pct":    1.23,
  "cavitation_band_energy_db": -34.5
}

```

Per-channel result (current [] and voltage [])

Each element in the `current` array (IA, IB, IC, IN) and `voltage` array (VA, VB, VC) has the following structure:

```

{
  "channel":      "IA",

```



```

"hl_rms":      12.453,
"rms_total":   12.510,
"thd_pct":     2.31,
"h3_ratio":    0.0082,
"valid_fund":  true,
"n_segments":  1,
"harmonics": [
  {
    "h":      1,
    "freq_hz": 59.998,
    "amplitude": 17.610,
    "amplitudeRms": 12.453,
    "phaseDeg": -12.4,
    "inPhase":  17.190,
    "quadrature": -3.782
  },
  ...
],
"spec_freq_hz": [ 0.0, 0.488, 0.977, ... ],
"spec_magnitude": [ 0.000012, 0.000015, ... ],
"fullband_peaks": [
  {
    "freq_hz":      59.998,
    "amplitude":     17.610,
    "snr_db":        48.2,
    "energy_ratio":  1.4142,
    "score":         0.987,
    "class":         "harmonic"
  },
  ...
],
"ni_peaks": [
  {
    "location":      126.4,
    "freq_hz":       61.716,
    "amplitude":      0.00813,
    "amplitude_db":   -41.8,
    "d2":            -0.00091,
    "snr_db":         9.2
  },
  ...
],
"bb_lo": { "freq_hz": 58.01, "amplitude": 0.0151, "ratio_h1": 0.00121 },
"bb_hi": { "freq_hz": 61.99, "amplitude": 0.0147, "ratio_h1": 0.00118 },
"ecc":    [ ... ],
"bpfo":   { "freq_hz": 106.6, "amplitude": 0.00101, "ratio_h1": 0.000081 },
"bpfi":   { ... },
"bsf":    { ... },
"ftf":    { ... }
}

```

The **fullband_peaks** array contains all significant peaks found across the full bandwidth up to Nyquist, each classified into one of:

Class	Description
harmonic	Integer multiple of f_0
broken_bar_sideband	Within tolerance of $f_0(1 \pm 2s)$
eccentricity_sideband	Within tolerance of $f_0 \pm kf_r$
bearing_bpfo	Near BPFO
bearing_bpfi	Near BPFI
bearing_bsf	Near BSF
bearing_ftf	Near FTF
slot_harmonic_zone	In the rotor slot harmonic region
high_freq_noise	Wideband high-frequency noise
unknown	Unclassified

advanced object (summary)

```
"advanced": {
  "hypotheses": {
    "broken_bar": 0.021,
    "eccentricity": 0.009,
    "bearing": 0.006,
    "supply_h3": 0.000,
    "current_unbalance": 0.031
  },
  "wavelet_family": "haar",
  "wavelet_levels": 7,
  "wpt_depth": 4,
  "stft": { "win_n": 4096, "hop_n": 2048,
    "time_s": [...], "freq_hz": [...],
    "mag_db": [...] },
  "f0_track": { "time_s": [...], "value": [...] },
  "slip_track": { "time_s": [...], "value": [...] },
  "thd_track": { "time_s": [...], "value": [...] },
  "h1_track": { "time_s": [...], "value": [...] },
  "sidebands": { "time_s": [...], "lower": [...], "upper": [...] },
  "wavelet_bands": [ { "level": 1, "energy": ..., "normalized_energy":
    ... } ],
  "wavelet_bands_db4": [ ... ],
  "wpt_nodes": [ { "depth": 4, "node": 3, "energy": ..., "
    freq_lo_hz": 4000, "freq_hi_hz": 6000, "label": "slot_harmonic" } ],
  "vi_power": {
    "time_s": [...],
    "active_power_h1": [...],
    "reactive_power_h1": [...],
    "power_factor_h1": [...],
    "z_h1_magnitude": [...],
    "z_h1_phase_deg": [...]
  },
  "startup": { "detected": false },
  "coupling": {
    "mean_h3_ratio": 0.082,
    "likely_supply_driven_h3": false
  }
}
```

decision object

```

"decision": {
  "severity": 0.034,
  "confidence": 0.021,
  "persistence": 0.08,
  "primary_fault": "none",
  "ranking": [
    { "type": "current_unbalance", "score": 0.031 },
    { "type": "broken_bar", "score": 0.021 },
    { "type": "eccentricity", "score": 0.009 }
  ],
  "summary": {
    "primary_fault": "none",
    "confidence": 0.021,
    "evidence": []
  },
  "wavelet_interpretation": {
    "transient_detected": false,
    "max_band_energy": 0.12,
    "dominant_band": 5
  },
  "stft_summary": {
    "avg_energy": 0.0031,
    "max_energy": 0.0420,
    "persistence": 0.08,
    "ridge_freq_hz": [ 60.0, 59.9, 60.1, ... ]
  },
  "ai_features": [ 0.021, 0.034, 0.08, ... ]
}

```

The **severity** score is a weighted combination:

$$\text{severity} = 0.28 s_{sb} + 0.22 s_{dwt} + 0.18 s_{thd} + 0.12 s_{unbal} + 0.10 \text{persist} + 0.10 s_{SNR} \quad (50)$$

The **ai_features** vector is a 20-element normalised feature vector $\in [0, 1]^{20}$ for direct ingestion by downstream ML classifiers, comprising: confidence, severity, persistence, THD mean/max, H3 ratio, unbalance, sideband level, wavelet band energy, DWT-db4 score, STFT energy metrics, power factor, impedance, H3 coupling ratio, and the five hypothesis scores.

Diagnostic Console Output (stderr)

All informational and diagnostic messages are written to `stderr` so they do not corrupt the JSON `stdout` output. Every message is prefixed with a tag in square brackets:

```

[DSP]    file=dsp.json  frequency_hz=ok  motor.rpm=ok  motor.poles=ok
[DSP]    f0=60.000 Hz
[DSP]    rpm=1785.000  poles=4
[CAL]    IA=3.0518e-03  IB=3.0480e-03  IC=3.0550e-03  IN=3.0518e-03  [A/
code]
[CAL]    VA=1.2207e-01  VB=1.2200e-01  VC=1.2210e-01  [V/code]
[FFT]    Current IA ...
[FFT]    Current IB ...
[FFT]    Current IC ...
[FFT]    Current IN ...
[f0]     Auto-detected f0=59.998 Hz (prior=60.000)
[FMEC]    Spectral fmech refined: fmech=29.755 Hz  RPM=1785.3  SNR=18.2 dB
[SLIP]    Spectral slip estimate: s=0.0082  RPM=1785.3

```

```
[NIPEAK] IA: 47 peaks (width=5, snr>=6.0dB)
[NIPEAK] IB: 45 peaks ...
```

Tag	Meaning
[DSP]	Fields parsed from <code>dsp.json</code> .
[CAL]	Per-channel gains loaded from <code>calibration.json</code> .
[WARN]	Non-fatal warnings (missing file, incomplete calibration, etc.).
[FFT]	Per-channel processing progress.
[f0]	f_0 auto-detection result and acceptance decision.
[FMEC]	Spectral f_r refinement result.
[SLIP]	Slip estimation from sideband search.
[NIPEAK]	NI Peak Detector peak count per channel.
[METER]	Fields parsed from <code>meter.json</code> .

Usage Examples

```
# Basic manual mode
./motor_fft_cpu tcp.raw 60 1785 4

# Auto-config from DSP JSON
./motor_fft_cpu tcp.raw --dsp dsp.json

# JSON output with calibration
./motor_fft_cpu tcp.raw --dsp dsp.json --cal calibration.json --json

# JSON + external bearings
./motor_fft_cpu tcp.raw --dsp dsp.json \
    --bearings bearings_weatherford_mg.json --json

# All spectra (current + voltage)
./motor_fft_cpu tcp.raw --dsp dsp.json --json \
    --spectrum ALL --include-voltage-spectrum

# Recommended for MCSA (current only, compact)
./motor_fft_cpu tcp.raw --dsp dsp.json --json --spectrum ALL_CURRENT

# Limit exported spectrum to 1200 Hz (analysis still full-band)
./motor_fft_cpu tcp.raw --dsp dsp.json --json \
    --spectrum ALL --max-freq 1200

# Use IB as reference channel
./motor_fft_cpu tcp.raw --dsp dsp.json --ref IB --json

# Override DSP params: 50 Hz, 1500 RPM, 2 poles
./motor_fft_cpu tcp.raw --dsp dsp.json 50 1500 2 --json

# Disable f0 auto-detection
./motor_fft_cpu tcp.raw --dsp dsp.json --no-autodetect-f0 --json
```

Important Notes

- The FFT is **always** computed at the full $N = 65536$ size (or the next power of two for short captures). The `-max-freq` option only clips the serialised JSON output — all fault

detection and full-band peak extraction operate on the full spectrum.

- Fault detection, sideband extraction, harmonic analysis, and `fullband_peaks` are all computed on the complete positive spectrum up to Nyquist ($f_s/2 = 16000$ Hz).
- The JSON spectrum payload can be large (~2–3 MB for `-spectrum ALL`). For production use with MongoDB, `-spectrum ALL_CURRENT` is recommended.
- This binary is designed for three use cases: (1) offline analysis, (2) integration with `diagnose.py`, and (3) storage in MongoDB.
- **Schema version 2:** the field `"schema_version": 2` identifies the current JSON output format. Downstream consumers should check this field for backward compatibility.